Environmental Imaging
To Investigate Subsurface Conditions
at the Umiak N16 Sump
Southwest of Tuktoyaktuk, Northwest Territories

prepared for

EnCana Corporation
Calgary, Alberta

directed by

KAVIK-AXYS Inc.
Calgary, Alberta

produced by

Essis Ltd.
Okotoks, Alberta

Respectfully submitted,
Essis Ltd.

Per:

Jason Bucko
Geophysics and Imaging

Approved:

Sean T. Clarke, P. Geoph.
Sr. Environmental Geophysicist

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ESSIS LTD.
Signature ____________________
Date ____________________
PERMIT NUMBER: P 6825
The Association of Professional Engineers, Geologists and Geophysicists of Alberta

October 17th, 2005
(1560 02)

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1. INTRODUCTION

This report describes an environmental imaging (EI) investigation carried out under contract to KAVIK-AXYS Inc. of Calgary, Alberta for EnCana Corporation of Calgary, Alberta. The survey was carried out on September 19th, 2005 at the Umiak N16 Sump, located southwest of Tuktoyaktuk, Northwest Territories (Figure 1).

The survey objectives were to investigate and delineate the extents of potential drilling wastes at the sump. This investigation was compared to a previous survey directed by KAVIK-AXYS Inc. (see Essis Project #1442, August 2004) to determine possible changes in subsurface ion-distributions.

The Geonics EM38 and EM31 were used to map the lateral extents of ion-contaminated regions using electromagnetic (EM) inductive forces. Two Trimble Pathfinder Power DGPS receivers provided positional control. Data were differentially corrected using Inuvik, Northwest Territories base station data.

All project work was carried out in accordance to discussions with Mr. Chris Revak of KAVIK-AXYS Inc. of Calgary, Alberta.

FIGURE 1
Site Location

2. ENVIRONMENTAL IMAGING (EI) - DATA ACQUISITION

EI refers to the imaging of the subsurface for anomalous environmental conditions. Various tools are applicable to this process, each of unique capability and purpose. This section describes the tools used to image environmental subsurface conditions on site.

Appendix A contains geophysical (Geonics EM38 and EM31) and DGPS positioning (Trimble Pathfinder Power) specifications. Appendix B contains a technical paper by the instrument manufacturer (Geonics Limited). This paper provides in-depth technical background on the applied geophysical technique (Frequency-Domain EM induction).

2.1 LATERAL CONDUCTIVITY (GEONICS EM38/31)

Appendix A contains Geonics EM38/31 instrument specifications. The Geonics EM38/31 use electromagnetic (EM) inductive forces to delineate subsurface conductivities. These instruments are digitally synchronized to differentially-corrected GPS (DGPS). This arrangement allows positioning and geophysical data to be stored simultaneously on handheld dataloggers. Data are acquired swiftly at high sampling density. EM38 readings were taken in the 'EM38 – Horizontal Coils at Ground Level' mode, approximating measurements of the initial 1.5 m of topsoil (Table 1). EM31 readings were taken in the 'EM31 – Horizontal Coils at Hip Level' mode, approximating measurements of the initial 5 m of topsoil (Table 1).

Please note that an inherent property of EM signals prevents deep penetration into conductive media. Tabulated depths of penetration (Table 1) are skindepths of the penetrating EM signal and are a function, in part, of overall soil conductivity (conductive soil reduces depth-of-penetration). The various coil orientations are not used to delineate the depth-extent of contamination but rather the depth-onset of contamination.

TABLE 1
EM38/31 Coil Orientations and Corresponding Effective Depths of Penetration

Coil Orientation	Effective Depth of Penetration
EM38 – Horizontal Coils at Ground Level	**1.5 m**
EM38 – Vertical Coils at Ground Level	1 m
EM31 – Vertical Coils at Ground Level	2 m
EM31 – Horizontal Coils at Hip Level	**5 m**
EM31 – Horizontal Coils at Ground Level	6 m

2.2 DGPS POSITIONING (TRIMBLE PATHFINDER POWER)

Appendix A contains DGPS instrument specifications. Positioning data were differentially corrected using Inuvik, Northwest Territories base station data. The moving DGPS receiver (Rover) was synchronized to geophysical instruments providing positioning control of geophysical measurements. Although this system can readily apply corrections in real-time, post-processing using specialized software is more accurate and avoids potential interference by real-time correcting radio signals. Sub-meter positioning accuracies are generally possible. Table 2 lists various DGPS survey parameters. Site features have been surveyed on site and serve as reference markings for follow-up work.

TABLE 2
DGPS Survey Parameters

Base Station UTM Coordinate:	N7577533/E560870/46.36 m (HAE)
Rover UTM Zone:	08
Final UTM Datum:	NAD27

3. ENVIRONMENTAL IMAGING (EI) - DATA PROCESSING AND PRESENTATION

EI data processing was initiated infield and completed at headquarters. Data processing was designed to enhance results by carefully applying specific mathematical algorithms. Such algorithms may filter unwanted data, enhance trend continuity or highlight otherwise unclear or invisible data.

3.1 Lateral Conductivity (Geonics EM38/31)

EM38/31 data were processed to moderate responses from buried metals and powerline interference. Additional data processing enhanced trend continuity and amplified hidden anomalies. Key data processing parameters are tabulated (Table 3) to enable future data processing repeatability. Two large-format ‘Lateral Conductivity Distribution (EM38/31) & Site Features’ maps were created (Maps 1/2 and 2/2) and are included in the clear plastic pouch accompanying this report. Reduced versions of both maps are bound into the report. The large-format maps detail site features better and offer finer EM-imaging than their smaller, bound counterparts.

The EM38 lateral conductivity image was overlain onto an aerial photograph (taken by Essis crew during the mobilization flight). As the aerial photograph exhibits focal distortion, the DGPS surveyed features on the maps should be used for referencing field locations. The aerial photograph underlay is bound in following the report.

Table 3

EM38/31 - Key Data Processing Parameters

Data Gridding:	Kriging (1 x 1)
	Moving Average (5, 5)
Data Imaging:	Non-linear, 4 – 20 mS/m
Image Enhancement:	Resolution enhancement to 150 dpi, Low-Pass @ 3

3.2 DGPS POSITIONING (TRIMBLE PATHFINDER POWER)

Positioning data were corrected for ionospheric effects. Specialized processing software by Trimble Navigation incorporated Inuvik, Northwest Territories base station data to differentially correct all rover positions. PDOP (Positional Dilution of Precision) was consistently below 2.5. PDOP is a statistical measure of positional accuracy related to satellite geometry relative to the receiver. In general, PDOP values in excess of five suggest somewhat more inaccurate positioning. Map UTM coordinates were projected in NAD27.

4. MAPS AND INTERPRETATION

Maps (1/2) and (2/2) are titled 'Lateral Conductivity Distribution (EM38/31) & Site Features'. The colour images on the maps represent the lateral conductivity distribution on site, also referred to as apparent conductivity. Apparent conductivity approximates the bulk conductivity from surface to skindepth of the primary EM field. Conductivity variations are generally caused by a combination of ionic contaminants (chlorides), natural soil salinity, buried metal (pipes, scrap, etc.), changes in soil type and saturation.

Metallic and ionic responses can often be distinguished by interpreting their characteristic behaviour. Such behaviour is a function of geometric coupling between the instrument and buried metal. Buried metal is easily identified when geometric coupling is maximized (for example, when surveying perpendicularly across pipelines). Consequently, buried metal is not always identified when coupling is minimized (for example, when surveying parallel to pipelines). Please note that while the EM38/31 respond to buried metal, these tools are not specifically designed to map buried metals.

The possibility of conductive media increases from dark blue (cold colours) to dark red (hot colours). Regions that have been surveyed (as revealed by the presence of geophysical measurement stations, fine crosses) and are not coloured (but appear white) are also presumed background. Colour (blue) was omitted to ease the presentation of black site features.

Background readings can vary between sites in response to varying soil types. Dry accumulations of sand or gravel typically yield low backgrounds. Naturally occurring ions, fertilized, fine-grained and/or saturated soils yield elevated backgrounds. Elevated backgrounds can sometimes match magnitudes typical of contaminated regions. Lab results of test hole locations from within anomalous regions (hot colours) should reveal the cause of anomalous responses. Similarly, lab results of test hole locations from outside anomalous regions (cold colours) should reveal the nature of background responses.

The large-format maps are scaled at 1:600. The small-format, bound maps are scaled at 1:790. Data are presented in UTM coordinates and projected in NAD27. The UTM grid coordinate system displays in units of metres (as opposed to degrees) and allows distance measurements to be read directly off the maps. UTM X and Y grid lines (the fine vertical and horizontal lines on the maps) are spaced 10 m apart. Site features have been DGPS surveyed and are included on the maps to provide reference control during follow-up work.

Aerial Photograph #1 presents an overlay of the EM31 image to an aerial photograph of the site (taken by Essis crew on September 19th, 2005 during a mobilization flight). As the aerial photograph exhibits focal distortion, the DGPS surveyed features on the UTM referenced maps should be used for finding field locations.

In general, Maps (1/2) and (2/2) reveal shallower and deeper conductivity data on site, respectively. Conductivity response magnitudes range from 4 to 20 mS/m. Such low background conductivities are typical of clean sands and silts commonly found in this region.

The initial survey conducted on this site was performed in the fall of 2004 (see Essis Project #1442, August 2004) to identify the extents of potential ionic contamination at the sump. At that time, elevated responses suggested ionic contaminants were concentrated along the central axis of the visible sump area. This current September 2005 survey was conducted to identify potential changes in contaminant extents and magnitudes. To facilitate comparison of site conditions, presentation parameters such as colour scale were maintained from the initial survey.

The central ‘topographic high’ outlines an area of the sump where fill material has been piled. The ‘topographic high’ is visible on the Aerial Photograph as the rectangular shaped lighter ground pattern. Water pooling somewhat obscures the northern extent of this feature. The actual dimensions of the sump may exceed this visible boundary.

Conductivity response trends on deeper imaging EM31 (Map 2/2) suggest potential drilling wastes are concentrated on the west side of the visible sump area, trending along its central axis. Shallower-imaging EM38 conductivities reveal similar response trends on the western side of the sump. Elevated responses also appear north and south of the topographic high on both maps, trending parallel to the rectangular feature. Elevated responses outside the topographic high are particularly evident on the EM38 map, suggesting predominantly shallow conductors. Generally wet conditions around the sump may have influenced these slightly elevated shallow responses.

An elevated EM38 response trend extending south from the southeast corner of the sump feature may identify a potential contaminant migration path. The suspect flow path may travel downslope from the sump and pool on the opposing side of a ridge, identified on the UTM referenced and aerial underlay maps.

Comparison of the current 2005-fall survey to the previous 2004-fall survey reveals similar conductivity trends within the topographic high. The EM31 instrument reported attenuated response magnitudes along the axis of the sump during the current survey.

The suspected contaminant migration path identified by the 2005-EM38 survey concurs with an elevated conductivity trend revealed during the 2004 survey. The 2004 trend exhibited comparatively reduced response magnitudes, matching those of suspected water saturated conditions on site.

Future soil coring should be strategically targeted to confirm suspected contaminated zones. Lab results from future borehole samples should be combined and correlated with geophysical data. The combined interpretation of all data ensures a comprehensive understanding of site conditions.

5. CONCLUSIONS AND RECOMMENDATIONS

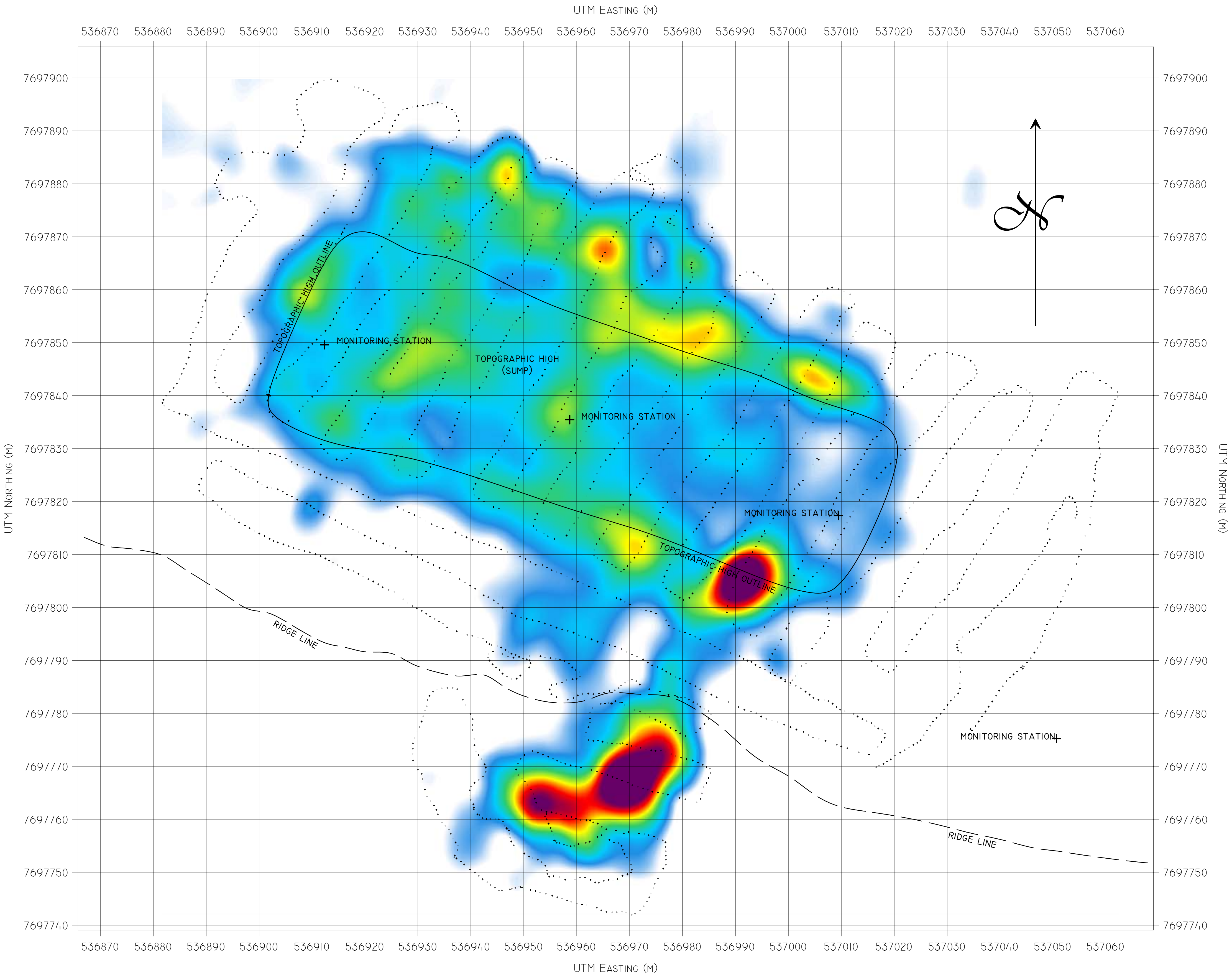
The survey objectives were to investigate and delineate potential drilling wastes at the Umiak N16 Sump. This investigation was compared to a previous survey directed by KAVIK-AXYS Inc. (see Essis Project #1442, August 2004) to determine possible changes in subsurface ion-distributions.

The Geonics EM38 and EM31 were used to map the lateral extents of ion-contaminated regions using electromagnetic (EM) inductive forces. Two Trimble Pathfinder Power DGPS receivers provided positional control. Data were differentially corrected using Inuvik, Northwest Territories base station data.

Geophysical findings from the current 2005 survey were compared to those from the previously conducted 2004 survey. The general sump location was identified by a pile of fill material, visible as a rectangular shaped topographic high on site. Lateral conductivity mapping revealed similar conductivity distributions as the previous year within the fill pile, where potential contaminants appeared concentrated along its central axis. Elevated responses appeared paralleling the northern and southern sides of the topographic high and may have been influenced by increased water saturation.

Potential ionic contaminant migration may have occurred towards the south from the southeast corner of the topographic high. Elevated response trends suggested the flow path extended over a ridge where potential contaminants may have pooled. Although this path was not readily apparent on last years results due to reduced response magnitudes, similar response trends suggest potential contaminant movement may have already been occurring.

Future soil sampling should be targeted to confirm the true nature of anomalous responses, with consideration give to target depth ranges. Lab results from future soil samples should be combined and correlated to all geophysical data. The combined interpretation of all data ensures a comprehensive understanding of site conditions. Site features have been DGPS surveyed to assist follow-up work. Caution is advised during subsequent sampling to avoid unidentified buried objects.



Site Location Map

Comments

The colour image reveals the 'Apparent Conductivity Distribution' on site. The term 'Apparent Conductivity' implies that conductivity measurements do not linearly relate to actual soil conductivities.

The presented EM data should be used qualitatively to select targets for vertical (depth) geophysical profiling or soil sampling. Geophysical results are only conclusive after correlation to soil sample data (ground-truthing). The depth response of the electromagnetic (EM) field varies from surface to a unique depth, the 'skindepth' of that particular EM signal. The skindepth of any EM signal is strongly influenced by overall soil conductivity. Since soil conductivity varies randomly, the apparent conductivity distribution does not represent data from any particular depth.

Posted features are surveyed using DGPS. A positional accuracy of several decimeters is generally possible. Positional accuracy diminishes near larger buildings and other satellite-obstructing features.

Technical Summary

Geophysical Specifications

Instrument:	Geonics EM38
Measured Quantities:	Quadrature in mS/m Inphase in ppt, Hs/Hp
Primary Field Source (Tx):	Self-Contained Dipole Tx
Receiver (Rx):	Self-Contained Dipole Rx
Intercoil Spacing:	1.0 m
Operating Frequency:	14.6 kHz
Conductivity Ranges:	10 - 1000 mS/m

Positioning (DGPS) Specifications

GPS Rover:	Trimble Pathfinder Power
GPS Base:	Trimble 4000ssi Dual
Data S/A Corrections:	Post-Processed Differential
Coordinate Projection:	Universal Transverse Mercator (UTM)
UTM Rover Zone:	8
Datum:	North American Datum 1927 (NAD27)
Base Coordinate:	SOPAC, Invuvik, NWT X: 560870E Y: 7577533N Elev.: 46.36 m (HAE)

Legend

⬭	Topographic High (Sump)
+	Monitoring Station
— — —	Ridge Line
········	Geophysical Measurement Stations

4 6 8 10 12 14 16 18 20

Apparent Conductivity
(mS/m)

Map 1/2	Scale = 1:790	September 2005	Job No.: 1560 02

Lateral Conductivity Distribution (EM38)
&
Site Features

Umiak N16 Sump

Client:
EnCana

Directed by:
Kavik-Axys Inc.

Produced by:
Essis

Approved: ____________

Created: ____________

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Date ____________

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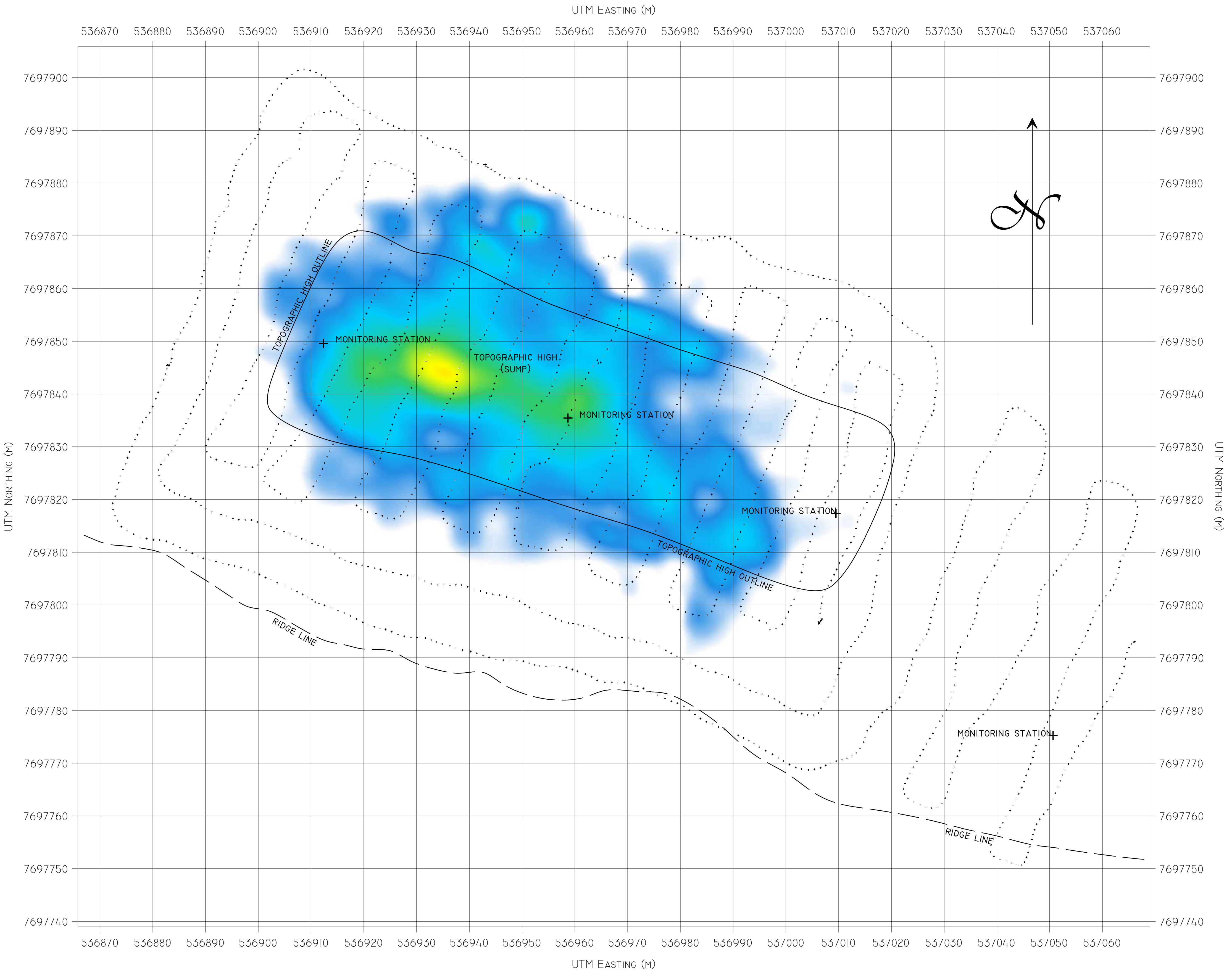
REDUCED VERSION



Site Location Map

Comments

The colour image reveals the 'Apparent Conductivity Distribution' on site. The term 'Apparent Conductivity' implies that conductivity measurements do not linearly relate to actual soil conductivities.

The presented EM data should be used qualitatively to select targets for vertical (depth) geophysical profiling or soil sampling. Geophysical results are only conclusive after correlation to soil sample data (ground-truthing). The depth response of the electromagnetic (EM) field varies from surface to a unique depth, the 'skindepth' of that particular EM signal. The skindepth of any EM signal is strongly influenced by overall soil conductivity. Since soil conductivity varies randomly, the apparent conductivity distribution does not represent data from any particular depth.

Posted features are surveyed using DGPS. A positional accuracy of several decimeters is generally possible. Positional accuracy diminishes near larger buildings and other satellite-obstructing features.

Technical Summary

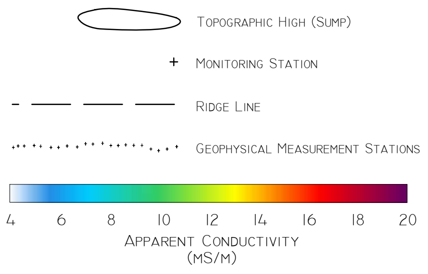
Geophysical Specifications

Instrument:	Geonics EM31
Measured Quantities:	Quadrature in mS/m Inphase in ppt, Hs/Hp
Primary Field Source (Tx):	Self-Contained Dipole Tx
Receiver (Rx):	Self-Contained Dipole Rx
Intercoil Spacing:	3.66 m
Operating Frequency:	9.8 kHz
Conductivity Ranges:	10 - 1000 mS/m

Positioning (DGPS) Specifications

GPS Rover:	Trimble Pathfinder Power
GPS Base:	Trimble 4000ssi Dual
Data S/A Corrections:	Post-Processed Differential
Coordinate Projection:	Universal Transverse Mercator (UTM)
UTM Rover Zone:	8
Datum:	North American Datum 1927 (NAD27)
Base Coordinate:	SOPAC, Invuvik, NWT X: 560870E Y: 7577533N Elev.: 46.36 m (HAE)

Legend



Map 2/2	Scale = 1:790	September 2005	Job No.: 1560 02

Lateral Conductivity Distribution (EM31)
&
Site Features

Umiak N16 Sump

Client:
EnCana.

Directed by:
KAVIK-AXYS Inc.

Produced by:
Essis

Approved: ____

Created: ____

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APPENDIX A

EM38
GROUND CONDUCTIVITY METER with INPHASE CHANNEL

The EM38 measures the conductivity of the ground directly in mS/m. The EM38 also measures the inphase component of response, which enables it to be used as a metal detector or, for poorly to moderately conductive material, a magnetic-susceptibility meter.

The depth to which the EM38 measures conductivity depends solely on the orientation of the instrument. When the instrument is upright in vertical-dipole mode (as shown), depth of measurement is approximately 1.5 m. With the instrument on its side, in horizontal-dipole mode, depth of measurement is approximately 0.75m. Readings are shown on digital meters which are mounted on the top and side of the EM38, for convenient reading regardless of the orientation of the instrument.

Using EM induction, the EM38 requires no contact with the ground. As a result, surveys can cover large areas rapidly. To survey at walking speed, a special handle allows the operator to carry the EM38 at ground level.

The EM38 is synchronized with DGPS instrumentation to simultaneously record positioning and geophysical measurements.

The EM38 is most commonly used during soil-salinity surveys, in both dry-land and irrigated areas. Compared to resistivity, EM38 surveys offer better lateral-resolution, and much faster coverage. The EM38 can be towed behind a skidoo or other vehicle to efficiently cover large terrain.

The EM38 can be applied to mapping ionic-contamination at a variety of site types including oil and gas, salt storage and waste-disposal. These applications make good use of the excellent spatial resolution of the EM38. The EM38 can also be used to identify shallow metallic objects.

**Technical information provided by Geonics Ltd., Mississauga, Canada. Phone (905) 670-9580*

EM38 SPECIFICATIONS

Measured Quantities:	Quadrature in mS/m Inphase in ppt, Hs/Hp	**Intercoil Spacing:** **Operating Frequency:**	1.0 m 14.6 kHz
Primary Field Source (Tx):	Self-Contained Dipole Tx	**Conductivity Ranges:**	10-1000 mS/m
Receiver (Rx):	Self-Contained Dipole Rx		

EM31
GROUND CONDUCTIVITY METER with INPHASE CHANNEL

The EM31 measures the conductivity of the ground directly in mS/m. The EM31 also measures the inphase component of response, which is useful for detecting ferrous and non-ferrous buried material.

Two meters on the front panel of the EM31 simultaneously display conductivity and inphase response. Readings can be taken at successive survey stations, or continuously along the survey line.

Using the inductive method, the EM31 operates without the ground contact required by conventional resistivity. As a result, surveys can be done over highly resistive material, such as gravel or asphalt, at the pace of a walk.

Compared to resistivity surveys, EM31 surveys yield detailed, continuous data, with better resolution of small changes in conductivity. This enables the EM31 to identify subtle changes in conductivity that can be caused by contamination, and to delimit affected areas with precision. Depth of exploration is about 6 m, which makes the EM31 suitable for many geotechnical and environmental applications.

The EM31 is synchronized with DGPS instrumentation to simultaneously record positioning and geophysical measurements. The EM31 can be towed behind a skidoo or other vehicle to efficiently cover large terrain.

The EM31 can be applied to mapping ionic-contamination at a variety of site types including oil and gas, salt storage and waste-disposal. The EM31 can map conductive contamination of soil and groundwater, and simultaneously detect buried metal, which makes this instrument the ideal tool for site assessment. It is most commonly used during soil-salinity surveys.

**Technical information provided by Geonics Ltd., Mississauga, Canada. Phone (905) 670-9580*

EM31 SPECIFICATIONS

Measured Quantities:	Quadrature in mS/m Inphase in ppt, Hs/Hp	**Intercoil Spacing:**	3.66 m
		Operating Frequency:	9.8 kHz
Primary Field Source (Tx):	Self-Contained Dipole Tx	**Conductivity Ranges:**	10-1000 mS/m
Receiver (Rx):	Self-Contained Dipole Rx		

GPS Pathfinder Systems

Versatile GIS data collection and maintenance

Key Features and Benefits

- **Total system solution**
- **Easy-to-use software**
- **High accuracy**
- **Real-time GIS data collection and maintenance**
- **Beacon and satellite differential capabilities**
- **Supports leading GIS database formats**
- **WAAS - capable**

Trimble's GPS Pathfinder® systems are a family of high-performance GIS data collection and maintenance products. These versatile systems offer a variety of software, data collector and GPS receiver options so you'll find one that is ideal for your needs. Powerful and easy-to-use, you can quickly collect quality data for utility, urban and natural resource databases. And as the demand for high-quality position and attribute information increases, these systems allow you to update existing GIS data—ensuring that your decisions are made with the most accurate, current and reliable data available.

Powerful, rapid, accurate data collection and maintenance

Easy-to-use Software

Time–saving field software is essential for productive GIS data collection and maintenance.

With Trimble's field software options, you can quickly and easily collect point, line and area features, along with their customized attribute information. Field software makes it easy to take existing data from your GIS into the field for verification and update of position and attribute information. With powerful navigation tools, the field software guides you to an existing feature or landmark using graphical display and textual messages.

Three options are available for collecting and maintaining quality data while out in the field.

Asset Surveyor® software runs on the revolutionary TSC1™ —a rugged, weatherproof, handheld data collector developed and built by Trimble.

TerraSync™ software operates on standard Pocket PC and Windows CE devices, providing the flexibility to choose from a wide range of devices depending on your requirments.

ASPEN® software runs on your pen or notebook computers, when extra processing power or storage is important for your field operations.

Real-time Receivers

Because you need immediate results, the GPS Pathfinder systems family includes Trimble's real-time differential GPS (DGPS) receivers. Real-time DGPS is key for relocating existing assets and for verifying that the correct feature is being updated.

The **GPS Pathfinder Pro XRS** system integrates GPS, real-time beacon, satellite differential and Wide Area Augmentation System (WAAS) capabilities.

The **GPS Pathfinder Pro XR** system integrates GPS, real-time beacon and WAAS capabilities.

The **GPS Pathfinder Power** receiver integrates GPS, real-time satellite differential, and WAAS capabilities into a single, lightweight unit.

Efficient Planning and Processing

The powerful GPS Pathfinder Office software allows you to quickly plan your data collection and maintenance work and process your field data for use in your GIS. Important functions such as data dictionary creation, data viewing and editing, and differential correction can all be completed with ease. GPS Pathfinder Office software can process real-time DGPS data, ensuring your data is of the highest quality before exporting it to many leading GIS packages.

With powerful field and office software and integrated real-time DGPS, the GPS Pathfinder systems family meets your GIS data collection and maintenance needs today and into the future.

GPS Pathfinder Systems

Versatile GIS data collection and maintenance

FEATURES AND OPTIONS

GPS Pathfinder Systems Standard Features

- GPS Pathfinder Office software
- Choice of GPS receiver
- Choice of field software
- Ergonomic backpack carrying system
- Rechargeable system batteries (provide 8 hours of field use)
- Battery charger and AC power supply

Optional Receiver Accessories

- Vehicle kit—includes cigarette lighter power adapter, quick release, 2 quick-release adapters and magnetic mount.
- GPS Pathfinder Centimeter Processing option
- Rangepole bipod system

Available Receivers and Standard Features

- GPS Pathfinder Pro XRS receiver
- GPS Pathfinder Pro XR receiver
- GPS Pathfinder Power receiver

- 12-channel GPS receiver
- EVEREST™ multipath rejection technology
- WAAS differential correction capabilities

Available Field Software

- Asset Surveyor software—for Trimble TSC1 data collector
- TerraSync software—for standard Pocket PC and Windows CE devices
- ASPEN software—for Windows-based pen or notebook computers

GPS PATHFINDER PRO XR AND XRS—RECEIVER & ANTENNA SPECIFICATIONS

GPS Pathfinder Pro XR

- Integrated GPS/Beacon receiver
- Integrated GPS/Beacon antenna
- RTCM input/output
- 3-meter antenna cable
- Base datalogging mode

GPS Pathfinder Pro XRS

- Integrated GPS/Beacon/Satellite Differential receiver
- Integrated GPS/Beacon/Satellite Differential antenna
- RTCM input/output
- 3-meter antenna cable
- Base datalogging mode

GPS Pathfinder Pro XR receiver

General	12 channel, L1/CA code tracking with carrier phase filtered measurements and multibit digitizer
Update rate	1 Hz
Power	6 Watts (max), 10 to 32 VDC
Accuracy (RMS) (Note A)	
MCORR400 differential correction	50 cm + 1 ppm on a second-by-second basis (horizontal) Submeter + 2 ppm on a second-by-second basis (vertical)
Carrier phase processing	30 cm + 5 ppm with 5 minutes tracking satellites 20 cm + 5 ppm with 10 minutes tracking satellites 10 cm + 5 ppm with 20 minutes tracking satellites 1 cm + 5 ppm with 45 minutes tracking satellites (with Centimeter Processing option)
RTCM beacon radio transmissions	Better than 1 meter (Note B)
Time to first fix	30 seconds (typical)
Size	11.1 cm x 5.1 cm x 19.5 cm (4.4" x 2.0" x 7.7")
Weight	0.76 kg (1.68 lbs)
Temperature	–30° C to +65° C (–22° F to +149° F) (operating) –40° C to +85° C (–40° F to +185° F) (storage)
Humidity	100% fully sealed
Casing	Dustproof, splashproof, shock resistant; sealed to 5psi

GPS Pathfinder Pro XRS receiver

Specifications for the Pro XRS receiver are the same as for the Pro XR receiver with the following exceptions:

Power	7 Watts (max), 10 to 32 VDC
Accuracy (RMS) (Note A)	
RTCM satellite differential correction	Better than 1 meter (Note B)

GPS Pathfinder Pro XR antenna

General	Right-hand, circular polarized; omnidirectional; hemispherical coverage
Size	15.5 cm diameter x 10.8 cm high (6.1" x 4.2")
Weight	0.49 kg (1.08 lbs)
Operating temp	–30° C to +65° C (–22° F to +149° F)
Storage temp	–40° C to +85° C (–40° F to +185° F)
Humidity	100% fully sealed
Casing	Dustproof, waterproof, shock resistant

GPS Pathfinder Pro XRS antenna

Specifications for the Pro XRS antenna are the same as for the Pro XR antenna with the following exceptions:

Size	15.5 cm diameter x 14 cm high (6.1" x 5.5")
Weight	0.55 kg (1.2 lbs)

GPS PATHFINDER POWER RECEIVER/ANTENNA SPECIFICATIONS

- Integrated GPS/Satellite Differential receiver and antenna
- RTCM input

General	12 channel, L1/CA code tracking with carrier phase filtered measurements.
Update rate	1 Hz
Power	3.1 Watts, 9 to 32 V
Accuracy (RMS) (Note A)	
MCORR400 differential correction	Submeter + 1 ppm on a second-by-second basis (horizontal) Submeter + 2 ppm on a second-by-second basis (vertical)
Carrier phase processing	30 cm + 5 ppm with 5 minutes tracking satellites 20 cm + 5 ppm with 10 minutes tracking satellites 10 cm + 5 ppm with 20 minutes tracking satellites 1 cm + 5 ppm with 45 minutes tracking satellites (with Centimeter Processing option)
RTCM satellite differential correction	Better than 1 meter (Note B)
Time to first fix	30 seconds (typical)
Size	15.2 cm diameter x 12.7 cm high (6" x 5")
Weight	0.625 kg (1.38 lbs)
Temperature	–30° C to +60° C (–22° F to +140° F) (operating) –40° C to +80° C (–40° F to +176° F) (storage)
Humidity	100% fully sealed
Casing	Fully sealed, dustproof, waterproof, shock resistant

TRIMBLE TSC1 DATA COLLECTOR SPECIFICATIONS

Logging memory	2 MB, memory extension through user accessible Type II ATA PC card slot (Note C)
Size	26.7 cm x 11.7 cm x 4.2 cm (10.5" x 4.6" x 1.65")
Weight	0.85 kg (1.875 lbs), including rechargeable Lithium Ion battery
Operating temp	–30° C to +65° C (–22° F to +149° F)
Storage temp	–30° C to +80° C (–22° F to +176° F)
Humidity	100% fully sealed against sand, dust and moisture, buoyant, waterproof against accidental immersion
Display	240 x 200 extended temperature graphics STN LCD display
Power	<1 Watt

(footnotes)

Note A: At least 5 satellites, PDOP ≤6, signal to noise ratio ≥6, satellite elevation mask at 15 degrees.

Note B: RTCM SC-104 standard format broadcast from a Trimble reference station.

Note C: Memory extension through user-accessible Type II PC card slot. 16 MB PCMCIA Data Cards are available (33050-16)

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The Data Dictionary Editor lets you build custom data collection menus to be uploaded to your data collector. Items can be quickly added, edited, and moved within the data dictionary. A graphic display makes the process far easier by helping you understand the structure of the data dictionary at a glance.

With Pathfinder Office's Batch Processor, you can automate your workflow. Data can be transferred from your data collector, differentially corrected, and exported to your GIS or CAD system as a single automated function. After your data has been processed, you can print it or plot it to scale using a pen plotter.

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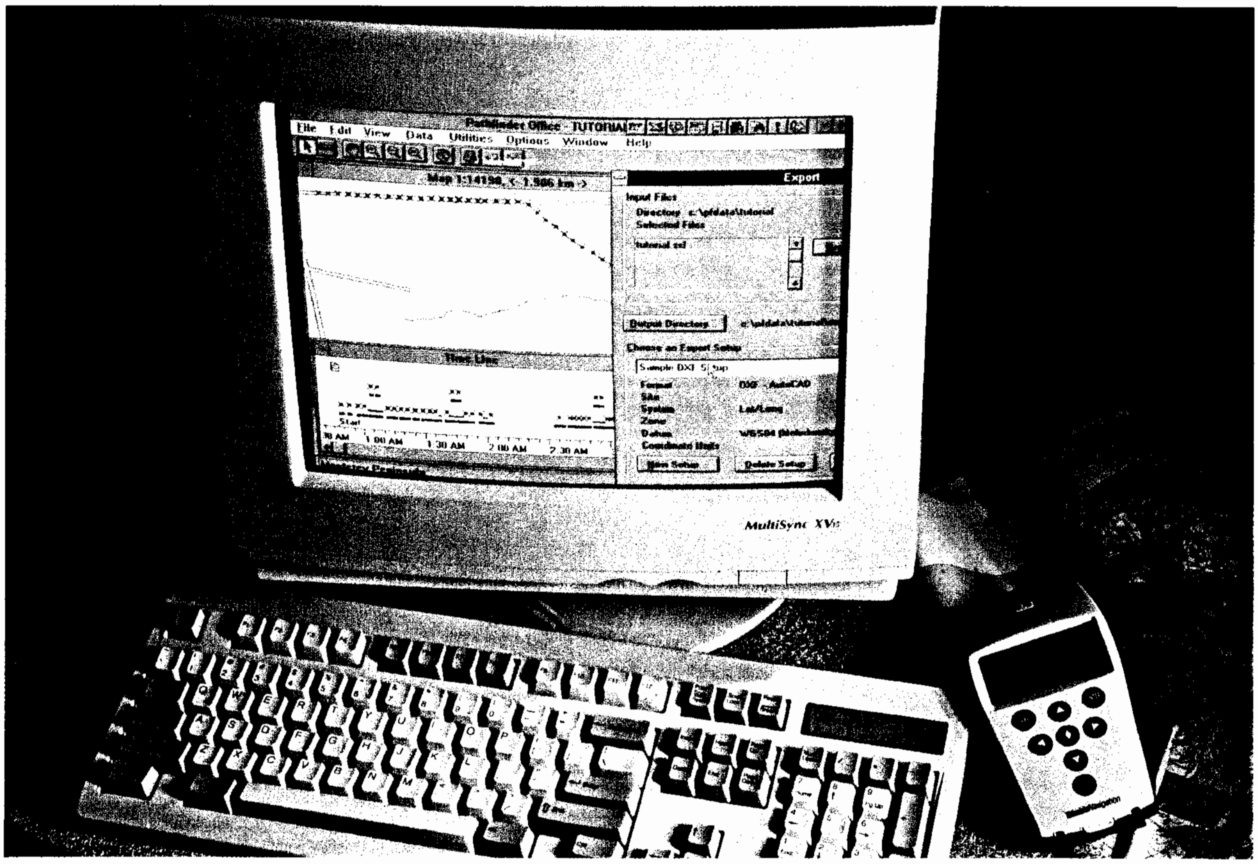
Pathfinder Office software sets a new standard for GPS planning and processing software. It provides all the tools needed to make your GPS data collection easy and successful.



Pathfinder Office Software

GPS Data Processing Software for Microsoft Windows

Pathfinder Office Software Capabilities

- Automated workflow with Batch Processor
- Export to most major GIS and CAD systems
- Carrier phase processing for high accuracy positions
- Display background maps behind GPS data
- Review and edit attribute data
- Import GIS files and take them to the field
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- Compatible with laser printers

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- Software CD
- Software Security Key (not included with upgrades)
- Activation Disk
- Printed Manuals
- On-Line Help
- Tutorial and Sample Files

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GEO-PC to Pathfinder Office upgrade	Part Number 31307-00
PFINDER to Pathfinder Office upgrade—current PFINDER Support Agreement required	Part Number 31308-00
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Computer Requirements

Recommended Configuration:

Computer:	Pentium microprocessor with 12 MB of RAM, a 100 MB hard drive, and a CD-ROM drive
Operating System:	Microsoft Windows 3.1 or later; Windows 95; or Windows NT 3.51 or later

Minimum Configuration:

Computer:	80486 microprocessor, 8 MB of RAM, and an 80 MB hard disk
Operating System:	Microsoft Windows 3.1 or later

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Trimble Navigation Limited
Mapping & GIS Systems Division
485 Potrero Avenue
Sunnyvale, CA 94086
1-800-545-7762 in North America
+1-408-481-8940
+1-408-481-7744 Fax
http://www.trimble.com

Trimble Navigation Europe Limited
Trimble House, Meridian Office Park
Osborn Way, Hook
Hampshire RG27 9HX U.K.
+44 1256-760150
+44 1256-760148 Fax

Trimble Navigation Singapore PTE Limited
300 Beach Road
#34-05 The Concourse
Singapore 199555
SINGAPORE
+65 296-2700
+65 296-8033 Fax

APPENDIX B

Table of Contents

GEONICS LIMITED
1745 Meyerside Drive, Unit 8, Mississauga, Ontario, Canada L5T 1C6 Tel.: (905) 670-9580 Fax: (905) 670-9204

Technical Note TN-6

ELECTROMAGNETIC TERRAIN CONDUCTIVITY MEASUREMENT at LOW INDUCTION NUMBERS

JD McNEILL

October, 1980

I. INTRODUCTION

The measurement of terrain resistivity to map geology has been utilized for over half a century. Several shortcomings, however, have prevented this technique from being widely accepted for engineering purposes. The first of these is that conventional galvanic resistivity surveys require a relatively large amount of manpower to execute and are thus expensive. Secondly, the actual value of resistivity itself is seldom diagnostic; it is the lateral or vertical variations of resistivity which form the basis of any interpretation. However the high cost of resistivity surveying generally means that fewer measurements are made than would be desirable, with the result that either (i) the survey area is not made large enough to establish a reasonable background against which the anomalous areas are to be delineated or (ii) the anomalous area itself is obscure and lacks definition.

An additional problem inherent to conventional resistivity techniques is that although the effective depth of exploration is determined by the selected inter-electrode spacing, resistive inhomogeneities which are small compared to this depth but which are located near the potential electrodes can cause a significant error in the measurement. Such fluctuations in the measured results are truly geological "noise" because it is not possible to determine the physical size, resistivity contrast, or location of the source. As a result of such inhomogeneities resistivity profiles carried out at constant interelectrode spacing tend to be noisy, limiting the resolution in resistivity that can be achieved, even though the instrumentation itself is capable of producing much higher accuracy.

It was an awareness of both the advantages of resistivity for engineering geophysical surveys and the disadvantages of conventional resistivity techniques that led Geonics Limited to examine the possibility of employing electromagnetic (inductive) techniques as an alternative for resistivity surveys. With the development of the EM31 and the EM34-3 it is now possible to map terrain conductivity virtually as fast as the operator(s) can walk; furthermore the sample volume is averaged in such a manner as to yield unexcelled resolution in conductivity.

These patented instruments have been designed to cover the range of depths generally useful for engineering geophysics; the EM31, one-man portable, has an effective depth of approximately 6 meters and the EM34-3, two-man portable, has stepwise selectable depths from 7.5 meters to a maximum of 60 meters.

Typical applications for the EM31 and EM34-3 instrumentation are:

(i) Delineating regions of permafrost (frozen pore water)
(ii) Locating gravel
(iii) Extending known gravel deposits
(iv) Mapping saline intrusions
(v) Detecting cavities in carbonate rocks
(vi) Mapping pollution plumes in groundwater
(vii) Mapped bedrock topography
(viii) Mapping terrain conductivity for electrical grounding
(ix) General geological mapping (soil types, fault and fracture zones, etc.)
(x) Archaeological exploration
(xi) Locating pipes (EM31) and metallic-type conductors

This technical note describes both the principles and the instrumentation employed to measure terrain conductivity using electromagnetic techniques at low induction numbers. For a detailed discussion of the concept of terrain resistivity/conductivity and of the various factors that control this parameter the reader is referred to Geonics Limited Technical Note "Electrical Conductivity of Soils and Rocks".

II. PRINCIPLE OF OPERATION

The application of electromagnetic techniques to the measurement of terrain resistivity, or more properly, conductivity* is not new and excellent descriptions of this technique are given in the literature [1], [2].

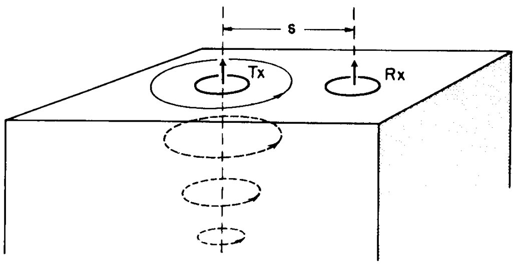

FIGURE 1. Induced current flow (homogeneous halfspace).

Consider Figure 1 in which a transmitter coil Tx energized with an alternating current at an audio frequency, is placed on the earth (assumed uniform) and a receiver coil Rx is located a short distance s away. The time-varying magnetic field arising from the alternating current in the transmitter coil induces very small currents in the earth. These currents generate a secondary magnetic field H_s which is sensed, together with the primary field, H_p, by the receiver coil.

In general this secondary magnetic field is a complicated function of the intercoil spacing s, the operating frequency, f, and the ground conductivity σ. Under certain constraints, technically defined as "operation at low values of induction number" (and discussed in detail in the appendix) the secondary magnetic field is a very simple function of these variables. These constraints are incorporated in the design of the EM31 and EM34-3 whence the secondary magnetic field is shown to be:

$$\frac{H_s}{H_p} \simeq \frac{i\omega\mu_o\sigma s^2}{4} \tag{1}$$

where H_s = secondary magnetic field at the receiver coil
H_p = primary magnetic field at the receiver coil
$\omega = 2\pi f$
f = frequency (Hz)
μ_o = permeability of free space
σ = ground conductivity (mho/m)
s = intercoil spacing (m)
$i = \sqrt{-1}$

The ratio of the secondary to the primary magnetic field is now linearly proportional to the terrain conductivity, a fact which makes it possible to construct a direct-reading, linear terrain conductivity meter by simply measuring this ratio. Given H_s/H_p the apparent conductivity indicated by the instrument is defined from equation (1) as

$$\sigma_a = \frac{4}{\omega\mu_o s^2}\left(\frac{H_s}{H_p}\right) \tag{2}$$

The MKS units of conductivity are the mho (Siemen) per meter or, more conveniently, the millimho per meter.

III. INSTRUMENTATION

The EM31 (shown in Figure 2) has an intercoil spacing of 3.7 meters, which yields an effective depth of exploration of about 6 meters. The instrument can also be operated on its side, in which

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*Conductivity is preferred with inductive techniques since the response is generally proportional to conductivity and inversely proportional to resistivity.

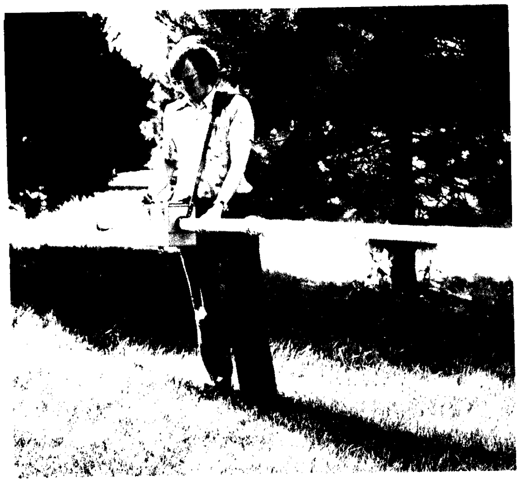


FIGURE 2. EM31 in field operation.

case as will be seen in Section IV., the effective depth of exploration is reduced to approximately 3 meters. The instrument is one-man portable and can be used either in "station-by-station" mode or read continuously. The presence of layering in the earth can be detected by raising the instrument and noting the readings as a function of instrument height. If the earth is two-layered the conductivity of both layers and the upper layer thickness can be resolved.

The EM34-3 which is two-man portable has the two coils flexibly connected (Figure 3). The intercoil spacing is measured electronically so that the receiver operator simply reads a meter to accurately set the coils to the correct spacing, which can be 10, 20, or 40 meters so as to directly vary the effective depth of exploration as shown in Table 1.

FIGURE 3. EM34-3 in field operation.

TABLE 1. Exploration depths for EM34-3 at various intercoil spacings

Intercoil Spacing (meters)	Exploration Depth (meters)	
	Horizontal Dipoles	Vertical Dipoles
10	7.5	15
20	15	30
40	30	60

To measure terrain conductivity the transmitter operator stops at the measurement station; the receiver operator moves the receiver coil backwards or forwards until his meter indicates correct intercoil spacing and he reads the terrain conductivity from a second meter. The procedure takes 10 to 20 seconds. The coils are normally carried with their planes vertical (horizontal dipole mode) since in this configuration the measurement is relatively insensitive to misalignment of the coils. In the event that the greater depth of penetration resulting when the two coils are in the vertical dipole mode is desired, more care must be taken with intercoil alignment. Because of the relatively short intercoil spacing correct alignment is usually not difficult to achieve.

Both instruments are calibrated to read terrain conductivity in millimhos per meter. To convert these readings to resistivity (in ohmmeters) one simply divides them into 1,000, i.e. 50 millimhos per meter is the equivalent of 20 ohmmeters.

IV. SURVEY TECHNIQUES AND INTERPRETATION

For either the EM31 or EM34-3 it can be shown that in a homogeneous or horizontally stratified earth the current flow is entirely horizontal. Furthermore under the constraints by which the instruments are designed the current flow at any point in the ground is independent of the current flow at any other point since the magnetic coupling between all current loops is negligible. Finally, under these constraints the depth of penetration is limited only by the intercoil spacing. We say that the depth of penetration is "source" or "geometry" limited rather than "skin depth" limited since it is now controlled by the fall-off with distance of the dipolar transmitter field. For this reason all dimensions are normalized with respect to the intercoil spacing in subsequent sections of this technical note.

IV. 1. Instrumental Response as a Function of Depth (Homogeneous Halfspace)

Consider a homogeneous halfspace on the surface of which is located an EM31 or an EM34-3 transmitter as shown in Figure 4. Fixing our attention on a thin layer of thickness dz at depth z (where z is the depth divided by the intercoil spacing s) it is possible to calculate the secondary magnetic field in the receiver coil arising from all of the current flow within this or any other horizontal thin layer. One can thus construct the function $\phi_V(z)$ shown in Figure 4 which describes the relative contribution to the secondary magnetic field arising from a thin layer at any depth z. We see from this figure that material located at a depth of approximately 0.4 s gives maximum contribution to the secondary magnetic field but that material at a depth of 1.5 s still contributes significantly. It is interesting to note that the ground at zero depth, i.e. the near surface material,

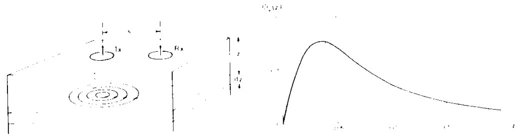


FIGURE 4. Relative response versus depth for vertical dipoles. $\phi_V(z)$ is the relative contribution to H_s from material in a thin layer dz located at (normalized) depth z.

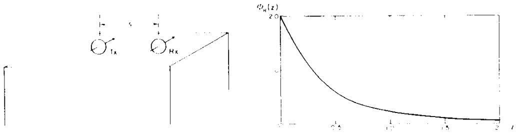


FIGURE 5. Relative response versus depth for horizontal dipoles

makes a very small contribution to the secondary magnetic field and therefore this coil configuration is insensitive to changes in near surface conductivity.

Figure 5 illustrates the function of Figure 4 for the case of both transmitter and receiver dipoles horizontal coplanar rather than vertical coplanar. For the coil configuration of Figure 5 (commonly used for the EM34-3 since it is less critical to intercoil alignment) the relative contribution from material near-surface is large and the response falls off monotonically with depth.

A comparison of the function ϕ for both coil configurations in Figure 6 emphasizes the different manner in which they respond to material at different depths. The difference is important since either instrument can be rolled over so that the vertical dipole transmitter/receiver geometry becomes a horizontal dipole transmitter/receiver geometry and vice versa. As will be seen later, this feature is useful in diagnosing and defining a layered earth. The figure also shows that for regions greater than one intercoil spacing in depth the vertical transmitter/receiver dipole gives approximately twice the relative contribution of the horizontal transmitter/receiver dipole.

To summarize, with either horizontal or vertical transmitter/receiver dipole orientation it is possible to construct a function which gives the relative response to the secondary magnetic field at the receiver from a thin layer of ground at any depth. That this is possible arises from the fact that (i) all current flow is horizontal and (ii) all current loops are independent of all other current loops. It should be noted that it is not possible to construct such functions for conventional resistivity techniques.

Finally, since as shown in Section II the definition of apparent conductivity is given in terms of the secondary magnetic field at the receiver, the functions in Figure 6 also give the relative contribution from material at different depths to the *apparent conductivity* indicated by the instrument meter. The integral of either function from zero to infinity gives the total secondary magnetic field at the receiver coil from a homogeneous halfspace which is directly related to the electrical conductivity of the halfspace by equation (1). It is therefore possible to state with great precision the relative influence of material at different depths to the indicated apparent conductivity.

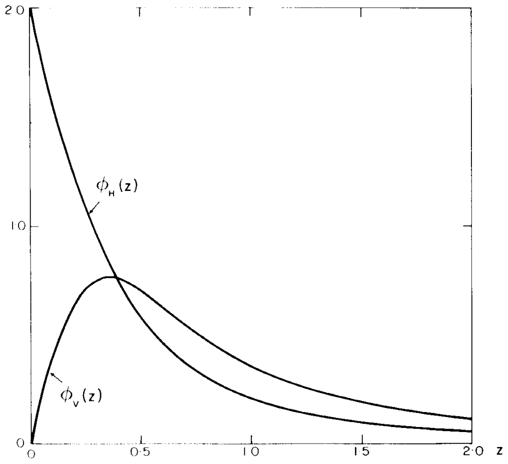


FIGURE 6. Comparison of relative responses for vertical and horizontal dipoles.

IV. 2. Multi-Layered Earth Response

The functions shown in Figure 6 are useful for describing the relative sensitivity of either of the two coil configurations to material at various depths. However a function derived from them is more useful for performing calculations. It is defined as the relative contribution to the secondary magnetic field or apparent conductivity from all material below a depth z and is given by

$$R_{\substack{V\\H}}(z) = \int_z^{\infty} \phi_{\substack{V\\H}}(z)dz \qquad (3)$$

Called the cumulative response, this function is illustrated in Figure 7 for vertical coplanar transmitter/receiver dipoles. The figure shows, for example, that for this configuration all material below a depth of two intercoil spacings yields a relative contribution of approximately 0.25 (i.e. 25%) to the secondary magnetic field at the receiver coil.

Suppose now that our homogeneous halfspace has a conductivity of 20 millimhos per meter (50 ohmmeters). The equipment having been calibrated according to equation (2), the output meter indicates 20 millimhos per meter. From Figure 7 we observed that the material below two intercoil spacings contributed 25% to the secondary magnetic field and therefore 25% to the indicated meter reading. Suppose that we replace this deep material with an infinitely resistive (zero conductivity) substance. Since we have reduced to zero the 25% that this material contributed to the meter reading the new reading will be 75% of 20, or 15 millimhos per meter. Conversely, if we leave all of the material below two intercoil spacings at 20

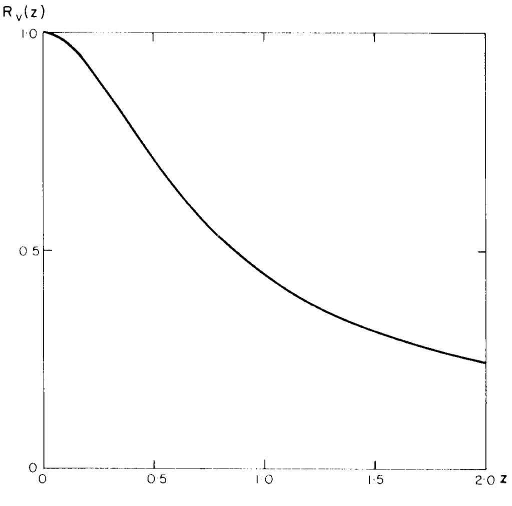


FIGURE 7. Cumulative response versus depth for vertical dipoles. $R_V(z)$ is the relative contribution to H_s from all material below a (normalized) depth z.

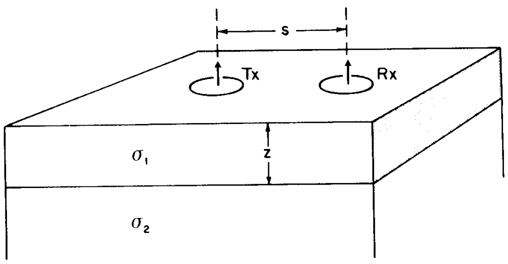


FIGURE 8. Two layer earth model.

millimhos per meter but make all material above two intercoil spacings infinitely resistive the meter reading will fall from the original 20 millimhos per meter for the homogeneous half space to 5 millimhos per meter, since, if all of the material below two intercoil spacings contributed 25% of the meter reading, all of the material above two intercoil spacings must contribute 75%; when removed the meter reading becomes 0.25 × 20 or 5 millimhos per meter.

From this example we see that there is a simple way to calculate the instrument reading on an arbitrarily layered earth as long as the intercoil spacing is much less than the skin depth in all of the layers. We simply add the contribution from each layer independently, weighted according to its conductivity and depth according to Figure 7. For example assume that we have a two-layer case as shown in Figure 8. The contribution from the upper layer is given by

$$\sigma_a = \sigma_1[1 - R_V(z)] \qquad (4a)$$

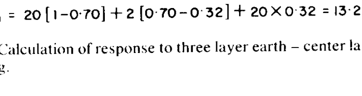
since all of the material below zero depth yields a relative contribution of unity or 100% to the meter reading. Conversely all of the material in the lower layer adds a contribution given by

$$\sigma_a = \sigma_2 R_V(z) \qquad (4b)$$

and the actual instrument reading will therefore be the sum of these two quantities

$$\sigma_a = \sigma_1[1 - R_V(z)] + \sigma_2 R_V(z) \qquad (5)$$

If the earth is three-layered as shown in Figure 9 the same procedure is employed to determine the instrumental response. In this example the calculations are performed for different middle layer thicknesses.



$$\sigma_a = \sigma_1[1 - R(z_1)] + \sigma_2[R(z_1) - R(z_2)] + \sigma_3 R(z_2)$$

$$z_2 = 1{\cdot}0, \quad \sigma_a = 20[1 - 0{\cdot}70] + 2[0{\cdot}70 - 0{\cdot}44] + 20 \times 0{\cdot}44 = 15{\cdot}3 \text{ mmho/m}$$

$$z_2 = 1{\cdot}5, \quad \sigma_a = 20[1 - 0{\cdot}70] + 2[0{\cdot}70 - 0{\cdot}32] + 20 \times 0{\cdot}32 = 13{\cdot}2 \text{ mmho/m}$$

FIGURE 9. Calculation of response to three layer earth – center layer thickness varying.

The ease with which such calculations are performed facilitates survey preparation and interpretation. It is sometimes possible to make advance estimates of the electrical properties of the materials to be encountered during a survey or, alternatively, once on-site the operator can obtain the same information from sample measurements of the different materials. The procedures outlined above are then employed to estimate the apparent conductivity measured under various terrain conditions. Examples of such calculations for the EM31 are shown in Figure 10. As is seen in the appendix the algebraic expressions for $\phi(z)$ and R(z) are very simple and are easily programmed on hand held calculators.

In Figure 10 the vertical dimensions are greatly exaggerated with respect to the horizontal dimensions. The question arises as to what degree of lateral uniformity is required before the earth can be considered as horizontally stratified or homogeneous. Survey experience indicates that if the ground conductivity does not significantly vary with horizontal distance within a radius of one intercoil spacing from the instrument the ground can be considered to be laterally uniform.

The above discussion referred to the use of vertical transmitter/receiver dipoles; it is equally possible to construct a cumulative response function for the horizontal coplanar dipole configuration and Figure 11 illustrates this function for both coil configurations. A comparison of the two curves illustrates that the vertical dipole mode of operation has approximately twice the effective exploration depth of the horizontal dipole mode.

IV. 3. Comparison with Conventional Resistivity Techniques

Many readers will be familiar with the two-layer curves employed to interpret data from conventional resistivity surveys using a Wenner array of four equally spaced electrodes. Using the techniques described in the previous section it is a simple matter to calculate two-layer curves for the electromagnetic technique; Figure 12 shows such curves for both the vertical and horizontal dipole configurations superimposed on standard Wenner curves. The general shape is similar but there are marked differences in detail. For vertical coplanar transmitter/receiver dipoles we see that when the substrate is the more resistive the response of the two systems is similar; however when the substrate is the more conductive the electromagnetic technique sees deeper in that the influence of the substrate, for a given conductivity contrast, is felt at smaller intercoil spacing than inter-electrode spacing. This is a general characteristic of electromagnetic systems which prefer to look through an insulator to a conductor rather than through a conductor to an insulator.

For the horizontal dipole configuration if the lower layer is the more resistive the effective exploration depth of the inductive technique is slightly less than the Wenner array; however, once again, in the case where the lower layer is the more conductive the exploration depth of the inductive technique is substantially greater.

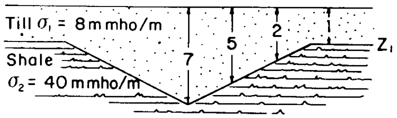
IV. 4. Resolution of Two-Layered Earth by Varying Intercoil Spacing

The principal advantage of the inductive electromagnetic technique over conventional resistivity lies in the speed and accuracy with which lateral changes of terrain conductivity can be measured. However this technique can also be used to measure the vertical variation of conductivity by expanding the intercoil spacing in a manner analogous to that in which the electrode spacing is expanded in conventional resistivity sounding techniques. The current state-of-the-art, however, is such that relatively few intercoil spacings can be employed; for example the EM34-3 can be operated with an intercoil spacing of 10, 20 or 40 meters. This feature is somewhat mitigated by the fact that the instruments can be used in either the vertical or horizontal dipole modes which, as shown in a previous section, exhibit different sensitivity to various depths thus yielding more information than would be available by simply using three spacings with one coil orientation.

To interpret a two-layer geometry the two-layer curves for both dipole configurations are superimposed on a common plot as shown

CROSS-SECTIONS

BURIED RIVER VALLEY

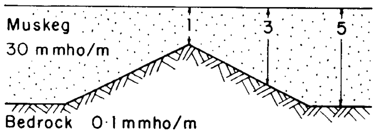


$$\frac{\sigma_a}{\sigma_1} = 1 - R(Z_1) + k_2 R(Z_1)$$

$$k_2 = \frac{\sigma_2}{\sigma_1} = \frac{40}{8} = 5$$

Z_1 (m)	σ_a (mmho/m)
1	32·6
2	26·9
5	18·6
7	16·0

BEDROCK HIGH

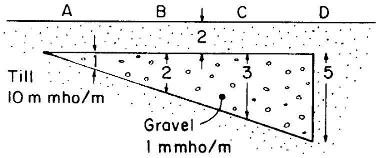


$$\frac{\sigma_a}{\sigma_1} = 1 - R(Z_1) + k_2 R(Z_1)$$

$$k_2 = \frac{\sigma_2}{\sigma_1} = \frac{0{\cdot}1}{30} = 0{\cdot}0033$$

Z_1 (m)	σ_a (mmho/m)
1	6·9
3	15·9
5	20·1

GRAVEL DEPOSIT



$$\frac{\sigma_a}{\sigma_1} = 1 - R(Z_1) + k_2[R(Z_1) - R(Z_2)] + k_3 R(Z_2)$$

$$k_2 = \frac{\sigma_2}{\sigma_1} = \frac{1}{10} = 0{\cdot}10$$

$$k_3 = \frac{\sigma_3}{\sigma_1} = 1{\cdot}00$$

station	σ_a (m mho/m)
A	8·9
B	8·2
C	7·7
D	6·9

FIGURE 10. EM31 calculated response across various geological features, using R(Z) corrected for instrument operation at waist (1 meter) height. Coil separation s = 3.67 meters.

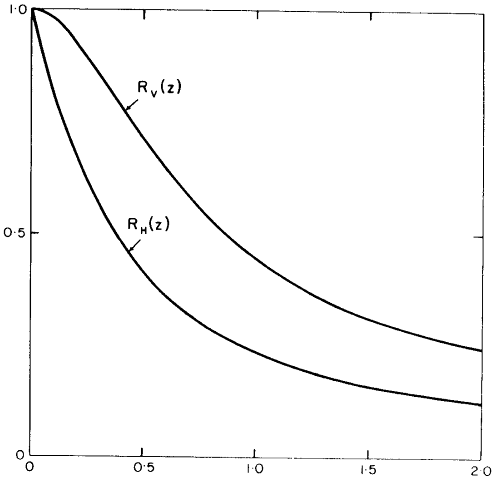


FIGURE 11. Cumulative response versus depth for vertical and horizontal dipoles.

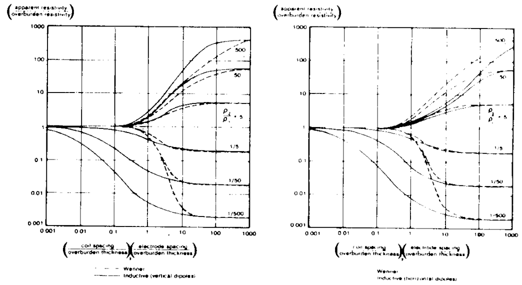


FIGURE 12. Comparison of Wenner array and inductive electromagnetic sounding curves for a two layer earth.

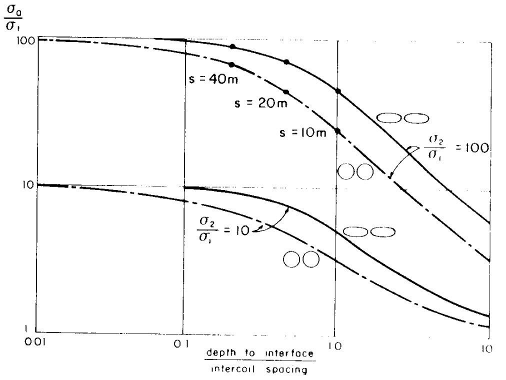
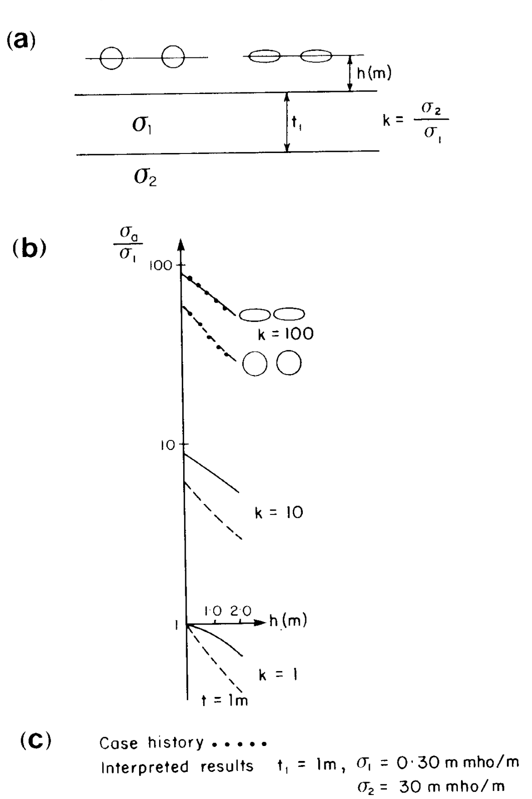


FIGURE 13. Two layer earth response curves (σ_2/σ_1 = 10,100; intercoil spacing varied). Dots indicate typical survey results.

in Figure 13. The six data points obtained by making measurements with two coil orientations and three intercoil spacings are plotted to the same scale on a piece of transparent paper and are translated vertically and horizontally on the two-layer curves to ascertain whether a satisfactory fit can be achieved. In the event that such a fit can be made, the earth does exhibit two-layer characteristics and the values of conductivity for both layers and the thickness of the upper layer are directly read off.

IV. 5. Resolution of Two-Layered Earth by Varying Instrument Height

In the case of the EM31 the intercoil spacing is rigidly fixed so that the technique described above is not available to analyse a layered earth. It is, however, possible to raise the instrument above the ground, measuring the apparent conductivity as a function of instrument height for both the vertical and horizontal dipole configurations. This has the effect of shifting the response curves of Figure 6 upwards through the various regions of the earth and the variation of apparent conductivity with height is therefore of diagnostic value in determining the nature of any layering. It is a straightforward matter to calculate the response of the instrument as a function of height for various two-layered earth geometries and typical curves are shown in Figure 14b. To use the curves one simply plots the measured apparent conductivity versus height for both coil configurations on a piece of transparent paper to the same scale as Figure 14b and shifts the plotted data vertically until good agreement is achieved with one of the curves, whereupon the two conductivities and the upper layer thickness are immediately determined as in the illustrated case history of Figure 14c.



Case history • • • • •

Interpreted results t_1 = 1m, σ_1 = 0·30 m mho/m

σ_2 = 30 m mho/m

FIGURE 14. Two layer earth response curves (σ_2/σ_1 = 1, 10, 100; instrument height varied). Dots are actual survey results.

In the event that the conductivity of either one of the two layers is known to be much less than the other, so that its contribution to the meter reading is negligible, it is simply necessary to lay the instrument on the ground, take a reading, lay it on its side, take a second reading, and from these two values one can immediately calculate the conductivity of the more conductive layer and the thickness of the upper layer.

V. ADVANTAGES AND DISADVANTAGES OF INDUCTIVE TERRAIN CONDUCTIVITY MEASUREMENTS

V. 1. Advantages

The advantages of the use of inductive electromagnetic techniques to measure terrain conductivity are as follows:

(i) *Excellent resolution in conductivity.* It was stated in Section I that a problem with conventional resistivity was that the presence of localized resistivity inhomogeneities near the potential electrodes caused large errors. If we examine the current flow in a homogeneous halfspace for the inductive technique described herein we realize that in the vicinity of the transmitter the current density is very high and we might expect the presence of a conductive inhomogeneity located here to have a large effect. However where the current density is high, the radius of the current loops is small and their distance from the receiver coil large, so that these loops do not couple well magnetically with the receiver. The effect of changing this current by varying the local conductivity is consequently negligible. The lateral extent of the volume of earth whose conductivity is sensed by the inductive technique is approximately the same as the vertical depth. The result is that small changes in conductivity, for example of the order of 5% or 10%, are easily and accurately measured.

(ii) *No current injection problems.* Since currents are magnetically induced in the earth, current injection problems encountered with conventional resistivity in materials such as gravel, bedrock, permafrost, snow and ice, etc., are not encountered with this type of instrumentation.

(iii) *Simple multi-layered earth calculations.* This matter is dealt with at length in Section IV.

(iv) *Easy, rapid measurements.* A problem with the conventional Wenner array is that in order to survey to an effective depth a the array must be 3a in length and the total length of wire required 4a, used in four sections. This presents many opportunities for snagging and breaking the wire. Furthermore each measurement requires insertion of four electrodes and relatively careful measurement of the inter-electrode spacing. These features are avoided with the inductive electromagnetic techniques and it is no exaggeration to say that a survey can often be carried out five to ten times faster using this technique. Indeed with either the EM31 or the EM34-3 it is usually possible under average terrain conditions to survey 5 to 7 line-kilometers a day with a station spacing of 25 or 50 meters.

V. 2. Disadvantages

As with all geophysical instruments, there are some limitations and disadvantages to the use of inductive electromagnetic techniques and these are as follows:

(i) *Limited dynamic range (1 – 1000 mmhos per meter).* At low values of terrain conductivity it becomes difficult to magnetically induce sufficient current in the ground to produce a detectable magnetic field at the receiver coil. Conversely at high values of conductivity the quadrature component of the received magnetic field is no longer linearly proportional to terrain conductivity as is shown in the appendix.

(ii) *Setting and maintaining the instrument zero.* Ideally in order to set the zero the instrument would be suspended in free space and the zero set there. The more acceptable alternative is to search out a region of very resistive ground, to accurately measure its conductivity using conventional techniques, and to set the instrumental zero at that location. This is the procedure which is actually followed.

It is necessary that this zero be accurately maintained over long periods of time and over the wide variations of temperature encountered during geophysical survey in various parts of the world. This produces tight constraints on the circuitry, with the result that the zero may be in error by up to ±0.2 mmhos per meter. Such an error would be negligible over the usual range of terrain conductivities; however in the event that measurements are being made on highly resistive ground the zero error can become significant.

(iii) *Limited Vertical Sounding Capability*. In theory it is possible to use a system such as the EM34-3 at a continuum of intercoil spacings to yield more information about electrical layering in the ground. To achieve a wide variety of inter-electrode spacings with conventional resistivity equipment is simple; in the case of the inductive electromagnetic technique the rapid fall-off of the magnetic field from the dipole transmitter introduces a serious dynamic range problem. In due course there will undoubtedly be instrumentation with a wider variety of spacings at the expense of additional complexity.

VI. CASE HISTORIES

This section describes several case histories obtained with the EM31 and the EM34. The surveys (i) illustrate the resolution in conductivity that can be achieved, (ii) compare the results obtained with conventional resistivity and (iii) illustrate the use of the latter for locating sand, gravel and conductive minerals, determining bedrock topography (including locating a buried river channel) and mapping the pollution plume from a land-fill site. In some cases the indicated conductivity has been converted to resistivity to facilitate comparison with conventional resistivity survey results.

Case History #1

Location: Mississauga, Ontario
Instrument: EM31
Application: Illustrates resolution and repeatability of EM31

For this case history a Rustrak chart recorder was used to monitor the output of an EM31. A line of length 200 meters was traversed in a field in both easterly and westerly directions. Figure 15 demonstrates that the instrument is resolving conductivity changes of less than 1 mmho/m (1% of full scale deflection) and that the repeatability is of the same order. In fact the repeatability is limited in this case by the resolving power of the chart recorder itself. It should furthermore be noted that the instrument is detecting spatial changes in conductivity of a few meters in length – compatible with the intercoil spacing of 3.7 meters.

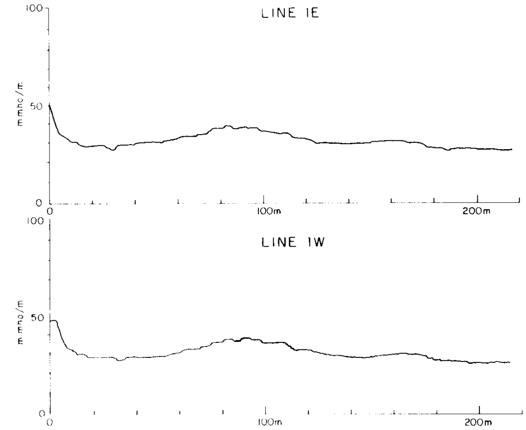


FIGURE 15.

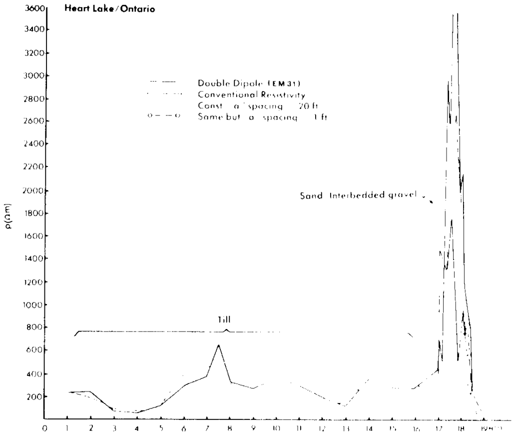


FIGURE 16. Test survey line – Heart Lake, Ont.

Case History #2

Location: Hearth Lake, Ontario
Instruments: EM31
Conventional resistivity apparatus
Application: Location of sand/gravel
Comparison of EM31 and conventional resistivity

In this survey a line 1900 ft. (580 meters) in length was surveyed with a measurement interval of 100 ft. (30 meters). The survey area was generally located on a buried esker, however the last few survey stations, 17 + 00 to 19 + 00, traversed a region of exposed sand and gravel (often occurring in the form of concretions) and over this portion of the line measurements were made every 10 ft. (3.0 meters).

The conventional resistivity profile was carried out using a Wenner array with an a spacing of 20 ft. (6.1 meters) except between stations 17 + 00 and 19 + 00 where the a spacing was reduced to 1 ft. (0.30 meters).

In general the correlation between the two sets of data is excellent, and demonstrates the ability of the EM31 to generate good quantitative data even in regions of low conductivity. Over the esker the EM31 was actually read continuously down the line – the data was recorded only at the 100 ft. intervals, with the exception of the reading at station 7 + 50 which was also recorded since it was noted that a conductivity low occurred there. Such an anomaly was, of course, missed by the conventional resistivity where measurements were only made every 100 ft.

Both sets of data become rather erratic between stations 17 + 00 and 19 + 00 as a result of the very rapid lateral changes in resistivity arising from the concreted material referred to above.

Case History #3

Location: Cavendish, Ontario.
Instrument: EM31
Application: Location of metallic type conductors

This survey line, of length 2000 ft. (610 meters), is located at a site

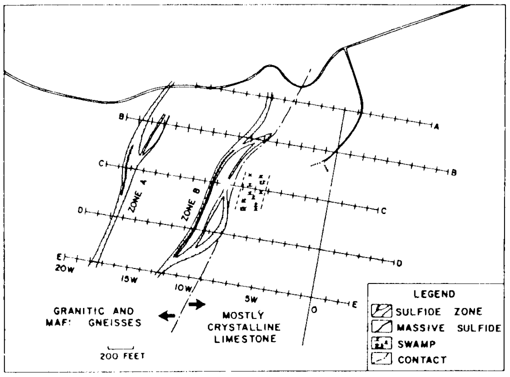


FIGURE 17. Geologic map of the Cavendish test site and the grid of traverse lines used in geophysical studies (after Ward et al [3]).

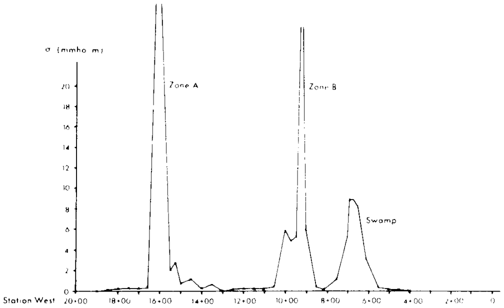


FIGURE 18. EM31 survey of Cavendish test range Line 'C'.

in Ontario which is often used by Canadian instrumentation manufacturers to test new electromagnetic geophysical equipment. The survey, along line C, illustrates response from both the swamp and the two zones of metallic mineralization. Although measurements were only taken every 50 ft. (15 meters) both zones are well delineated and when such high responses are encountered localization to within a few meters is quickly and easily carried out.

Inasmuch as the EM31 and EM34-3 were designed to map terrain conductivity at the conductivity levels encountered in typical soils both instruments are extremely sensitive electromagnetic detectors. For example on the most sensitive scale, full scale deflection for the EM31 is 800 ppm of the primary magnetic field and for the EM34-3 it is 3800 ppm. Such sensitivity makes either instrument useful for detecting metallic type conductors at what are very low conductivity levels by normal standards.

Case History #4

Location: Mississauga, Ontario
Instruments: EM31, EM34
Application: Determination of bedrock topography

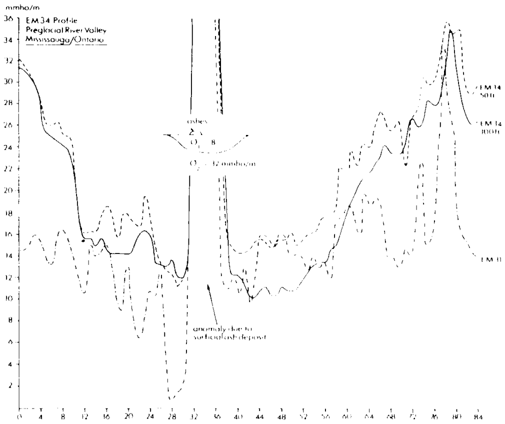


FIGURE 19. EM31 and EM34 survey line over preglacial river valley, Mississauga, Ontario.

Total line length for this survey was 8400 ft. (2600 meters) and measurements were made every 100 ft. (30 meters) with both the EM31 and the EM34 – an earlier version of the EM34-3 which had two intercoil spacings vis. 100 ft. (30 meters) and 50 ft. (15 meters). The survey was performed to outline the cross-sectional profile of a buried preglacial river valley whose existence had been suggested from water-well data. At either intercoil spacing the time required for the EM34 profile was 1-1/2 hours, resulting in approximately one survey measurement per minute – including the time to walk the 100 feet between measurement stations. The time taken for the subsequent EM31 survey was similar.

Typical bedrock conductivity in the area is approximately 30 mmho/m, whereas an average value for the conductivity of the infilling glacial till is of the order of 8 to 12 mmho/m. Thus the EM34 at either intercoil spacing yields approximately 30 mmho/m at the valley edges where the overburden is thin and 12 to 14 mmho/m at the valley centre. The EM31 yields values of 14 to 18 mmho/m at the valley edges (slightly affected by the presence of bedrock) and approximately 10 mmho/m at the valley centre. The interpreted depth of the valley, based on the model shown in the figure, is approximately 120 feet (36 meters) which is in reasonable agreement with the water-well data value of 150 feet (45 meters), bearing in mind that the three sets of data show that a two-layer model is an over simplification.

The conductivity high which occurs between stations 32 and 38 results from a very large pile of waste furnace ash lying on the surface.

Case History #5

Location: Camp Borden, Ontario
Instruments: EM31, EM34
Conventional resistivity apparatus
Application: Mapping groundwater salinity
Comparison of EM34 and conventional resistivity

Geophysical surveys were carried out over a sanitary landfill site using, in addition to other instruments, an EM31, EM34 and conventional resistivity [4]. The survey results in the accompanying figures illustrate the good agreement between these techniques and also indicate the reduction in survey time achieved using inductive electromagnetic techniques. Particularly interesting are the vertical variations in resistivity as shown by the EM31 at 3.7 m intercoil spacing and the EM34 at 15 and 30 m spacing.

VII. SUMMARY

This technical note describes in detail the principles of mapping the electrical conductivity of the ground using magnetically induced

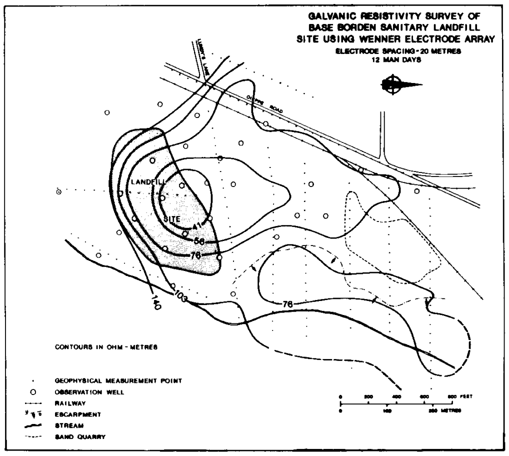


FIGURE 20(a).

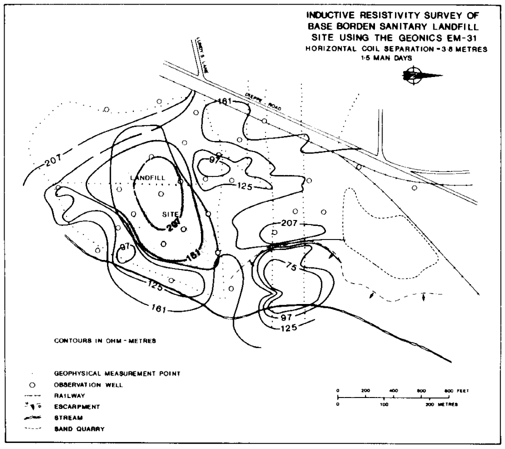


FIGURE 20(b).

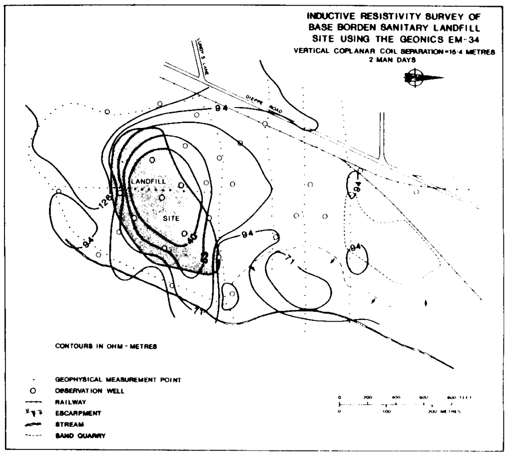


FIGURE 20(c).

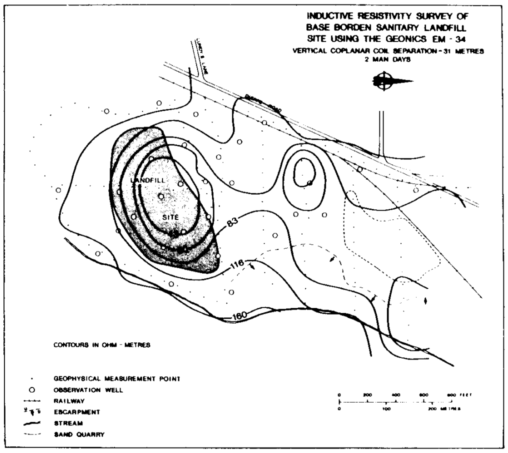


FIGURE 20(d).

currents at low frequencies. It has been shown that certain advantages can be derived from working at low values of induction number. Amongst these are excellent resolution in conductivity, a substantial reduction in man-hours necessary to carry out a conductivity survey and a simplification in the calculation of layered earth response.

Two points should be kept constantly in mind when performing surveys of this type to map geology. The first is that these instruments map only the electrical conductivity. If the conductivity does not vary significantly with the geological environment, or if parameters other than the geology also influence the conductivity, the survey results may be difficult to interpret.

The second point is that measurement of terrain conductivity, like any other geophysical measurement, must begin and end with geology. Such measurements are only an aid to help visualize geological conditions which cannot be seen. It is always necessary to interpret geophysical data against known geology from out-crops, boreholes, or any other such "bench marks". Geophysical measurements can be very effective by allowing interpolation between such sources, or extrapolation away from them. However in every case knowledge derived from geophysical measurements must be eventually reconfirmed against known geological conditions.

BIBLIOGRAPHY

(1) Keller, G.V., Frischknecht, F.C. Electrical Methods in Geophysical Prospecting. Pergamon Press 1966.
(2) Wait, J.R. 1962. A Note on the Electromagnetic Response of a Stratified Earth. Geophysics V.27, pp 382-85.
(3) Ward, S.H.; Pridmore, D.F.; Rijol, Glenn W.E. Multispectral Electromagnetic Exploration for Sulphides. Geophysics Vol. 39 No. 5 p. 666. 1974.
(4) Survey carried out by Dr. J. Greenhouse, University of Waterloo, Waterloo, Ontario.

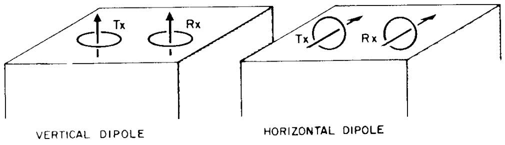


FIGURE AI. Vertical and horizontal dipole coil configurations.

APPENDIX: Theory of Operation at Low Induction Numbers

Consider the two coil configurations shown in Figure AI. In each case the transmitter coil is energized with alternating current at a frequency f Hertz. The measured quantity is the ratio of the secondary magnetic field H_s at the receiver when both coils are lying on the surface of the homogeneous half-space of conductivity σ to the primary magnetic field H_p in the absence of the half-space (i.e. as if the coils were in free space). The spacing between the coils is s meters.

The field ratios for vertical and horizontal dipole configurations are given by equations (1) and (2) respectively.

$$\left(\frac{H_s}{H_p}\right)_V = \frac{2}{(\gamma s)^2}\{9 - [9 + 9\gamma s + 4(\gamma s)^2 + (\gamma s)^3]e^{-\gamma s}\} \quad (1)$$

$$\left(\frac{H_s}{H_p}\right)_H = 2\left[1 - \frac{3}{(\gamma s)^2} + [3 + 3\gamma s + (\gamma s)^2]\frac{e^{-\gamma s}}{(\gamma s)^2}\right] \quad (2)$$

where $\gamma = \sqrt{i\omega\mu_0\sigma}$
$\omega = 2\pi f$
f = frequency (Hz)
μ_0 = permeability of free space
$i = \sqrt{-1}$.

These expressions are complicated functions of the variable γs which is in turn a reasonably complicated (complex) function of frequency and conductivity. However, as will be shown below, under certain conditions they can be greatly simplified.

A well known characteristic of a homogeneous half-space is the electrical skin depth δ, which is defined as the distance in the half-space that a propagating plane wave has travelled when its amplitude has been attenuated to 1/e of the amplitude at the surface. The skin depth is given by

$$\delta = \sqrt{\frac{2}{\omega\mu_0\sigma}} = \frac{\sqrt{2i}}{\gamma} \quad (3)$$

and therefore

$$\gamma s = \sqrt{2i}\,\frac{s}{\delta} \quad (4)$$

The ratio s/δ, the intercoil spacing divided by the skin depth, is defined as the induction number B, whereupon

$$\gamma s = \sqrt{2i}\,B \quad (5)$$

Now if B is much less than unity (ie $\gamma s \ll 1$) it is a simple matter to show that the field ratios of equations (1) and (2) reduce to the simple expression

$$\left(\frac{H_s}{H_p}\right)_V \simeq \left(\frac{H_s}{H_p}\right)_H \simeq \frac{iB^2}{2} = \frac{i\omega\mu_0\sigma s^2}{4} \quad (6)$$

which is the equation given in Section II.

The magnitude of the secondary magnetic field is now directly proportional to the ground conductivity and the phase of the secondary magnetic field leads the primary magnetic field by 90°.

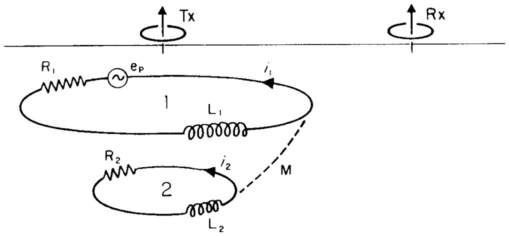


FIGURE AII. Electrical model for vertical dipoles.

To make B much less than unity we see that we must make s very much less than δ and thus

$$\omega \ll \frac{2}{\mu_0\sigma s^2} \quad (7)$$

That is, having decided on a value for s (which fixes the effective depth of penetration under the condition $B \ll 1$), the maximum probable ground conductivity is estimated and the operating frequency is chosen so that equation (7) is always satisfied.

The apparent conductivity which the instrument reads is then defined by

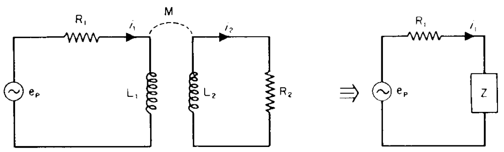
$$\sigma_a = \frac{4}{\omega\mu_0 s^2}\left(\frac{H_s}{H_p}\right)_{\substack{\text{quadrature}\\ \text{component}}} \quad (8)$$

To examine the reasons for this simplification let us focus our attention on the vertical dipole coil configuration shown in Figure AII since symmetry makes this configuration the simplest to understand.

Consider current loop 1. The primary emf e_p causing this current to flow is given (through Faraday's law) by the time rate of change of the primary magnetic flux from the transmitter through this loop. Three impedances cause the current to be limited. These arise from (i) the electrical resistance R_1 of the loop, (ii) the fact that the current i_1 generates its own magnetic field which causes a time-varying secondary magnetic flux through the loop (self-inductance, L_1), and (iii) the fact that all other current loops such as i_2 generate their own magnetic fields which in turn cause a time-varying magnetic flux to link with loop 1 (mutual-inductance, M).

The equivalent circuit for this configuration is easily derived from elementary circuit theory with the result shown in Figure AIII.

The complex impedance Z incorporates all of the affects of magnetic coupling between current loop 1 and any other current loop 2. We see from this expression that Z can be made arbitrarily small by reducing $\omega = 2\pi f$, the operating frequency. When Z is thus



$$Z = i\omega L_1 + \frac{\omega^2 M^2}{R_2 + i\omega L_2}$$

$$i_1 = \frac{e_P}{R_1 + Z}$$

FIGURE AIII. Equivalent circuit for model of Figure AII.

made much smaller than R_1 the current flow in loop 1 is simply given by

$$i_1 = \frac{e_p}{R_1} = \frac{i\omega\phi_p}{R_1} = i\omega\phi_p G_1 \tag{9}$$

where ϕ_p = primary flux linking loop 1
G_1 = conductance of loop 1 ($G_1 = 1/R_1$)
$i = \sqrt{-1}$

We see that the magnitude of the current is linearly proportional to the loop conductance and furthermore that the phase of the current leads the primary flux by 90°. Since the secondary magnetic field at the receiver from current i_1 is in phase with and directly proportional to i_1 it too will be directly proportional to G and will lead the primary flux by 90°. Thus

$$\left(\frac{H_s}{H_p}\right) \propto i\omega G_1 \tag{10}$$

which has the same dependence on frequency and conductance as equation (6). We infer therefore, that the condition $B \ll 1$ is equivalent to stating that for all current loops that affect the receiver output the operating frequency is so low that we can ignore any magnetic coupling between the loops. Thus the current that flows in any loop is (i) completely independent of the current that flows in any other loop since they are not magnetically coupled and (ii) is only a function of the primary magnetic flux linking that loop and of the local ground conductivity.

The lack of interaction between current loops is of great importance in simplifying the data reduction procedures. Of equally great significance is the fact that for any value of B and for any orientation of a magnetic dipole (or indeed of any magnetic source) over either a uniform halfspace or a horizontally stratified earth it can be shown that all current flow is horizontal. That this is the case for a vertical dipole is easy to see from symmetry; for a horizontal dipole it is less evident but equally true. Thus, in a horizontally layered earth no current crosses an interface which is fortunate since, if it did, changing either of the conductivities would, by virtue of refraction of the current, change the direction of the current as it flowed from one medium to the other.

If no current flow crosses an interface and if there is no magnetic coupling between current loops, changing the conductivity of any one of the layers of a horizontally stratified earth will not alter the geometry of the current flow. Varying the conductivity of any layer will proportionately vary only the magnitude of the current in that layer. To calculate the resultant magnetic field at the surface of a horizontally layered earth it is simply necessary to calculate the independent contribution from each layer, which is a function of its depth and conductivity, and to sum all the contributions.

The functions $\phi(z)$ and $R(z)$ discussed in Section II define the relative influence of current flow as a function of depth. Their derivation is involved and will not be given here. The resultant expressions are, however, simple and easily programmed into hand calculators:

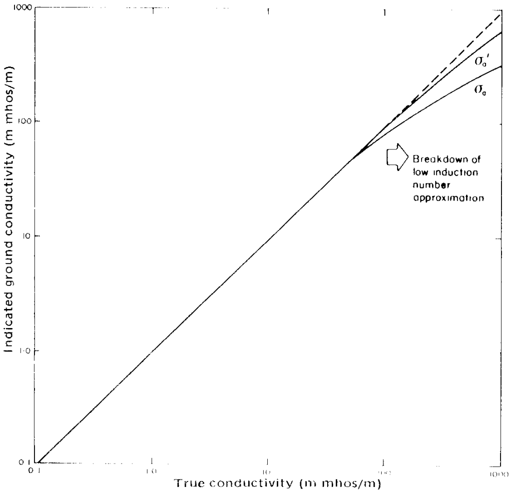


FIGURE AIV. Plot of indicated conductivity for EM31 versus true (homogeneous half-space) conductivity for both vertical (σ_a) and horizontal (σ_a') dipoles.

$$\phi_V(z) = \frac{4z}{(4z^2 + 1)^{3/2}} \tag{11}$$

$$\phi_H(z) = 2 - \frac{4z}{(4z^2 + 1)^{1/2}} \tag{12}$$

$$R_V(z) = \frac{1}{(4z^2 + 1)^{1/2}} \tag{13}$$

$$R_H(z) = (4z^2 + 1)^{1/2} - 2z \tag{14}$$

where z is the depth divided by the intercoil spacing.

Finally it should be noted that for a given frequency and intercoil spacing as the terrain conductivity increases the approximation of equation (6) eventually breaks down and the instrumental output is no longer proportional to terrain conductivity. This effect is illustrated in Figure AIV, which plots apparent (indicated) conductivity against true (homogeneous halfspace) conductivity for both vertical and horizontal transmitter/receiver dipoles for the operating parameters of the EM31. As would be expected the horizontal dipoles exhibit linearity to greater values of conductivity as a result of the reduced depth of penetration in this configuration.